

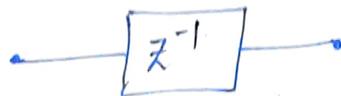
# Z - TRANSFORM;

→ Purpose: we want to convert Laplace domain to Z domain (engineer)

→ we want to make complex domain to polar domain

→ For storage purpose we go for Z transform.

→ we use shift register →



Invention of Z-Transform is done in the yr 1965.

It is invented by Raggazini & Jader.

→ generalisation of D.T.F.T.

→ discrete counter part of L.T.

$$x(t) = e^{st} \Rightarrow y(t) = e^{st} H(s).$$

$$x[n] = z^n \Rightarrow y[n] = z^n H(z).$$

$$z = re^{j\omega}.$$

$$X(z) = \sum_{n=-\infty}^{\infty} x[n] z^{-n}$$

$$x[n] \xleftrightarrow{z \cdot T} X(z)$$

$$X(re^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n] r^{-n} e^{-j\omega n}$$

Application of Z.T:

→ Sampled data controls

Jury's  
Lyapunov's

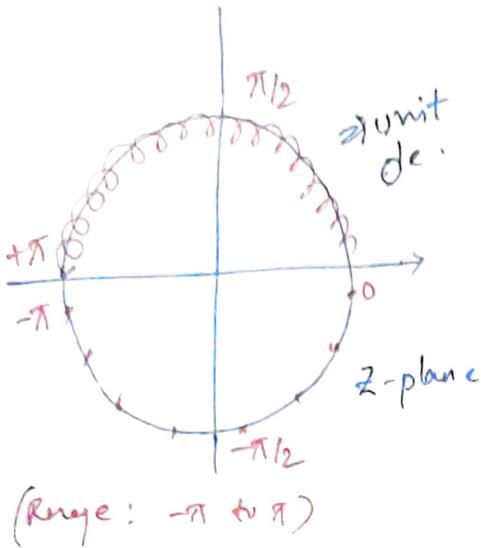
→ methods

→ D.F. Design  
(Digital filter)

Z-transform

$$X(z) = \text{F.T} \left\{ x[n] z^{-n} \right\}$$

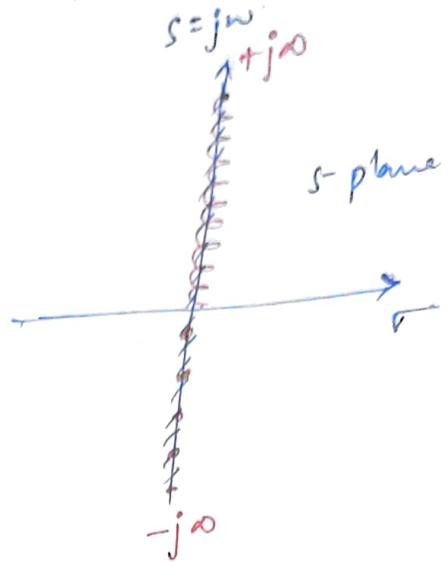
$$z=1 \Rightarrow z \cdot T = \text{D.T.F.T}$$



Laplace transform

$$X(s) = \text{F.T} \left\{ x(t) e^{-st} \right\}$$

$$\sigma=0 \Rightarrow \text{L.T} = \text{C.T.F.T}$$



- positive part of  $j\omega$  axis corresponds to upper part of the  
( $\omega$  varies from 0 to  $\pi$ )
- negative part of  $j\omega$  axis corresponds to lower half of  
the circle ( $\omega$  varies from  $-\pi$  to 0)

→  $\boxed{z = e^{sT_s}}$  → rel'n b/w L.T. & Z.T.

$$x(t) = e^{st}$$

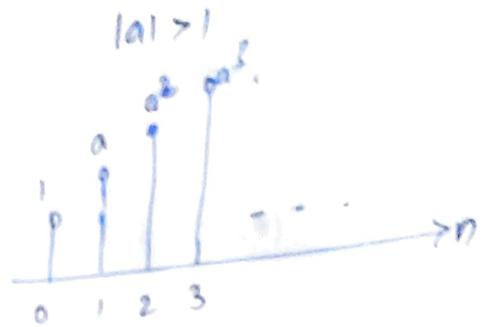
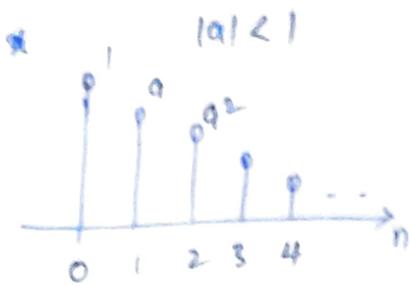
$$x[nT_s] = e^{snT_s}$$

$$x[n] = z^n$$

- Always Z.T is done with reference to Laplace transform
- Always D.T.F.T is done with ref to C.T.F.T.

z-transform of standard signals:

1.  $x_1[n] = a^n u[n]$



$$X_1(z) = \sum_{n=0}^{\infty} a^n z^{-n} = \sum_{n=0}^{\infty} (a z^{-1})^n$$

$$= 1 + a z^{-1} + (a z^{-1})^2 + \dots$$

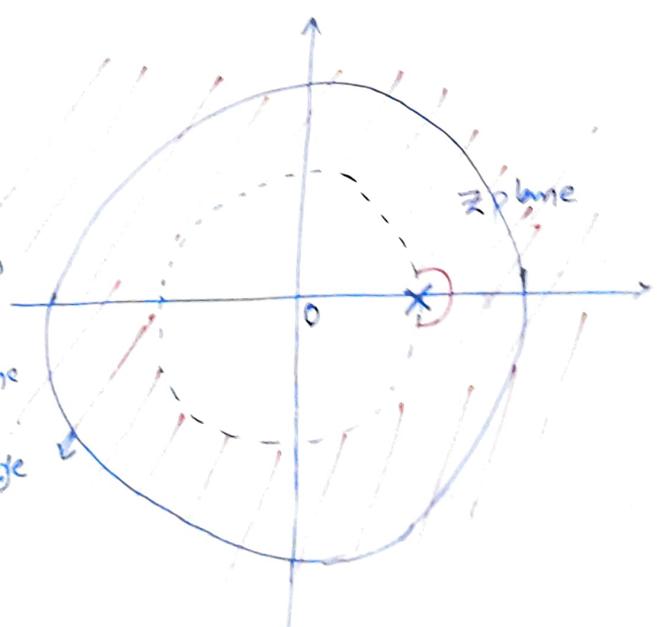
$$X_1(z) = \frac{1}{1 - a z^{-1}} ; \quad |a z^{-1}| < 1$$

$$\left| \frac{a}{z} \right| < 1$$

$$\Rightarrow \underbrace{|z| > |a|}_{\text{ROC}}$$

$$a^n u[n] \longleftrightarrow \frac{1}{1 - a z^{-1}} \quad (\text{or}) \quad \frac{z}{z - a} ; \quad |z| > |a|$$

1 pole at  $z = a$   
 $\Rightarrow$  one pole  
 if 2 poles are there we draw two poles  
 $\rightarrow$  Unit circle is the reference  
 unit circle



→ If the given sequence is right sided ROC is outside

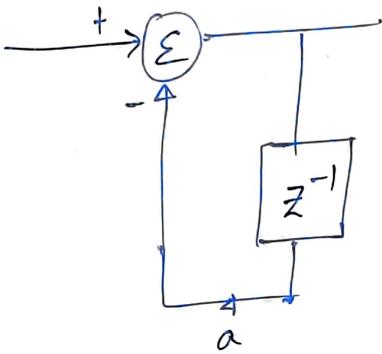
→ " " " " " " left sided " " inside.

R.O.C

Right sided  $\Rightarrow$  outside

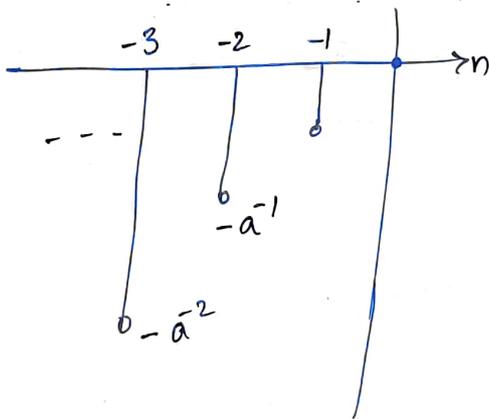
left sided  $\Rightarrow$  Inside.

→ To implement or generate  $a^n u[n]$  sequence sample we use the following digital cbt



2.  $x_2[n] = -a^n u[-n-1] = -a^n ; n \leq -1$

↓  
Signal is left sided.



$$X_2(z) = \sum_{n=-\infty}^{-1} -a^n z^{-n}$$

put  $-n = m$ .

$$= - \sum_{m=0}^{\infty} a^{-m} \cdot z^m$$

$$= - \sum_{m=1}^{\infty} (a^{-1}z)^m = - \left[ (a^{-1}z) + (a^{-1}z)^2 + \dots \right]$$

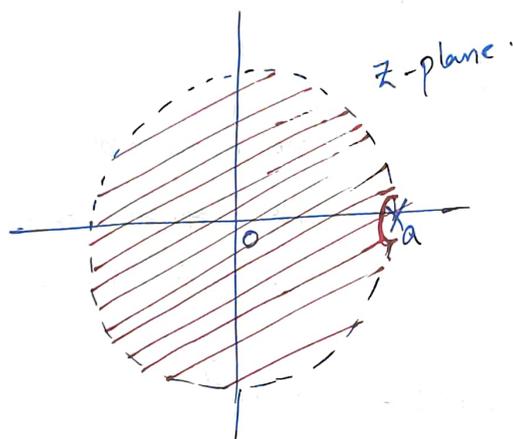
$$= \frac{-a^{-1}z}{1-a^{-1}z} ; |a^{-1}z| < 1$$

$$\Rightarrow |z| < |a|$$

$$\therefore -a^n u[-n-1] \xleftrightarrow{z.T} \frac{1}{1-a\bar{z}^{-1}} \quad (or) \quad \frac{z}{z-a} ; |z| < |a|$$

→ from the previous signal we observe that two is different signals but identical transforms with different ROC's.

ROC



→ when we apply inverse ~~z~~ z-transform

→ if the given sequence is left sided use

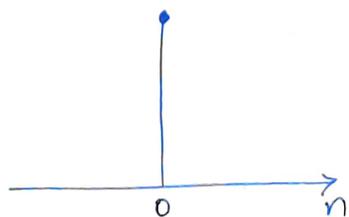
$$-a^n u[-n-1]$$

→ if the given sequence is right sided use  $a^n u[n]$ .

(\*)

$$x[n] = \delta[n]$$

$$X(z) = 1$$



ROC: entire z-plane

•  $S[n-1] \xrightarrow{z^{-1}} z^{-1} (1) \rightarrow$  positive sided  $S[n-1]$

ROC: entire  $z$ -plane except  $z=0$  ;  $|z| > 0$

for all +ve sided signal  $z=0$  will be exception

•  $S[n+1] \xrightarrow{z^{-1}} z^{-1(-1)} = z \rightarrow$  negative sided  $S[n+1]$

ROC: entire  $z$ -plane except at  $z=\infty$  (or)  $|z| < \infty$

for all -ve sided signal  $z=\infty$  will be exception

• Pg-146.

$n \rightarrow -2 \ -1 \ 0 \ 1$

$2 \cdot 1 \cdot 3$  (a)  $x_1[n] = \{ 1, 2, 3, -1 \}$

ROC:  $0 < |z| < \infty$

( $\because$  we have both positive & negative side impulse exception for ROC is

$z=0$  &  $z=\infty$ .)

(b)  $x_2[n] = \left(\frac{1}{2}\right)^n [u[n] - u[n-10]]$

$0 \leq n \leq 9$

It is a finite length signal.

i.e., ROC is entire  $z$  plane except zero.

ROC:  $|z| > 0$ .

(c)  $x_3[n] = \left[ \left(\frac{1}{2}\right)^n + \left(\frac{3}{4}\right)^n \right] u[n-10]$

$n \geq 10$

It is infinite length

The two poles are  $\frac{1}{2}$  &  $\frac{3}{4}$



7.1.2  $x[n] = (-1)^n u[n] + a^n u[-n-n_0]$ .

to find 'a' & 'n<sub>0</sub>' to get the ROC as  $1 < |z| < 2$

sg  $x[n] = (-1)^n u[n] + a^n u[-n-n_0]$

↓

$|z| > 1$

$|z| > 1$

↓

$|z| < |a|$

$|a| = 2$

$a = \pm 2$  & "n<sub>0</sub>" can be any value

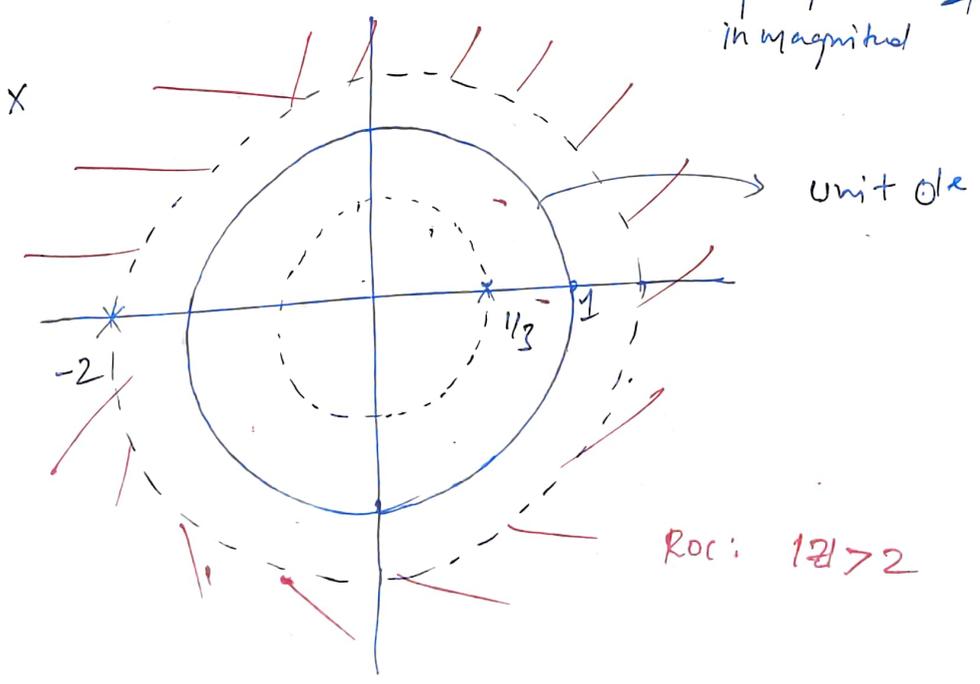
Rational Z.T concept

Eg:  $X(z) = \frac{1}{(1 - \frac{1}{3}z^{-1})(1 + 2z^{-1})} = \frac{1/7}{(1 - \frac{1}{3}z^{-1})} + \frac{6/7}{(1 + 2z^{-1})}$

Poles at  $z = \frac{1}{3}$  &  $-2$ . ( $a = -2$ )

Case (i):  $x[n]$  is right-sided.

largest pole in magnitude  $\Rightarrow |-2| = 2$

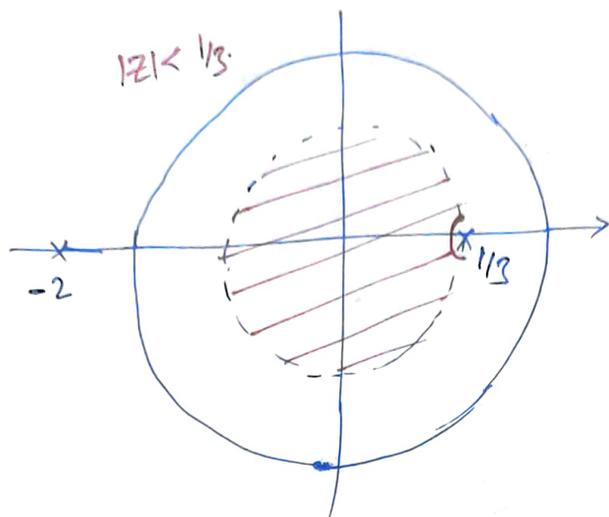


It is Causal

$$x[n] = \frac{1}{7} \left(\frac{1}{3}\right)^n u[n] + \frac{6}{7} (-2)^n u[n]$$

↓ right sided
↓ right sided.

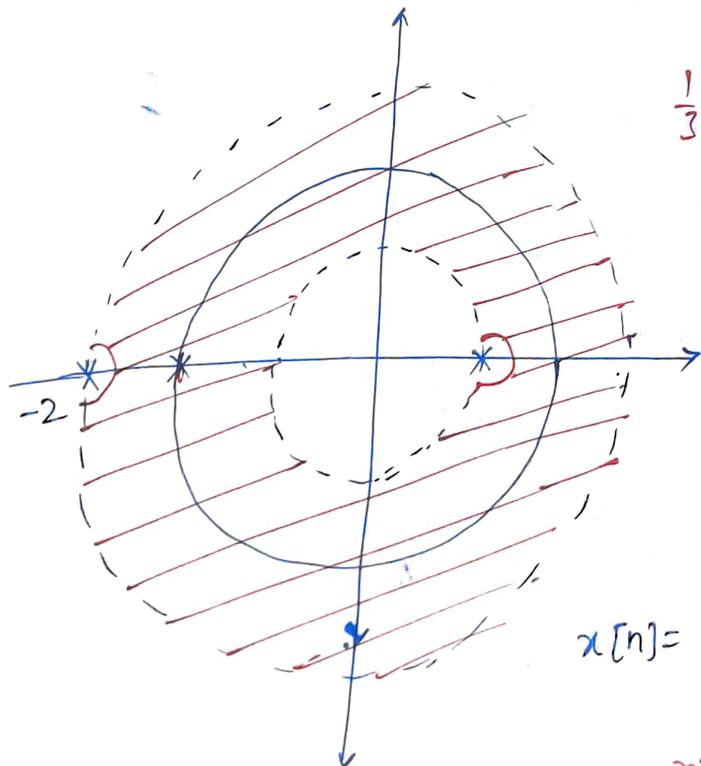
Case-ii.  $x[n]$  is left sided.



$$x[n] = \frac{1}{7} \left(\frac{1}{3}\right)^n [-u[-n-1]] + \frac{6}{7} (-2)^n [-u[-n-1]]$$

↓ left sided
↓ left sided

Case-3:  $x[n]$  is two sided.



$$x[n] = \frac{1}{7} \left(\frac{1}{3}\right)^n u[n] + \frac{6}{7} (-2)^n [-u[-n-1]]$$

↓ right sided
↓ left sided.

(DIRDO).

$$X(z) = \frac{z^2}{(z - e^{j\pi/2})(z - e^{-j\pi/2})}$$

$$= \frac{z^2}{(z+j)(z-j)} = \frac{2z^2}{z^2+1}$$

$\therefore X(1) = 1$

right sided  $\Rightarrow |z| > |\pm j|$   
 ROC:  $|z| > 1$

Ans: d

$$y(n) = \left(-\frac{1}{3}\right)^n u[-n-2].$$

We have  $x[n-n_0] \xleftrightarrow{z^{-T}} z^{-n_0} X(z).$   
 $x[n] \xleftrightarrow{z^T} X(z).$

Now  $x[n] = a^n u[-(n+1)].$

$$x[n+1] = a^{n+1} u[-(n+2)].$$

$$y(n) = \left(-\frac{1}{3}\right)^n u[-n-2]$$

$$z^T \left( = \left[-\frac{1}{3}\right]^{\overbrace{n+1-1}^{n_0=-1}} u[-(n+2)] \right)$$

$$Y(z) = -3 \left[ \frac{-z^{-(-1)}}{1 + \frac{1}{3}z^{-1}} \right]; |z| < \frac{1}{3}$$

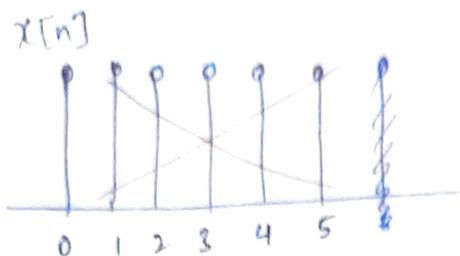
$$x[n] = \begin{cases} 1; & 0 \leq n \leq 5 \\ 0; & \text{else where} \end{cases}$$

$$g(n) = x(n) - x(n-1)$$

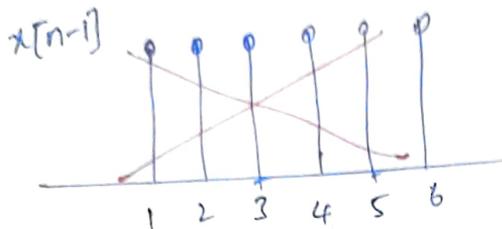
$$x(n) \xrightarrow{z^{-1}} X(z)$$

$$G(z) = X(z) - X(z)z^{-1}$$

q



$$x[n-n_0] \longleftrightarrow z^{-n_0} X(z)$$

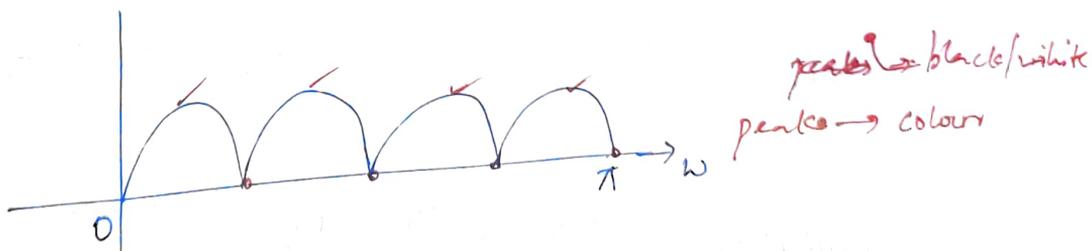


$$g[n] = x[n] - x[n-1]$$

$$z^{-1} \downarrow = \delta[n] - \delta[n-1]$$

$$G(z) = 1 - z^{-1} ; |z| > 0$$

↓  
 → The application of this system is Comb filter. It is used in T.V which separates colour & black/white component



⊙  $X(z) = \frac{z^3 - 2z}{z - 2}$  ;  $x[n]$  is left sided.

sd  $X(z) = \frac{z(z^2 - 2)}{z - 2}$

$$z^{-2} z^3 - 2z (z^2 + 2z)$$

$$\frac{z^3 - 2z^2}{z^2 - 2z}$$

$$0 + 2z^2 - 2z$$

$$\frac{2z^2 - 4z}{z^2 - 2z}$$

$$+ 2z$$

$$X(z) = \frac{z^2 + 2z}{z-2} + \frac{2z}{z-2}$$

$$(\because \delta[n-n_0] \xleftrightarrow{z \cdot T} z^{-n_0})$$

$$X(z) = \frac{z^2 + 2z}{z-2} + \frac{2z}{z-2}$$

$\swarrow \text{I} \cdot z \cdot T$        $\swarrow \text{I} \cdot z \cdot T$        $\swarrow \text{I} \cdot z \cdot T$

$$x[n] = \delta[n+2] + 2\delta[n+1] + 2(2)^n [-u[-n-1]]$$

$\downarrow$   
 left sided.

Method - 2

$$X(z) = \frac{z^3}{z-2} - \frac{2z}{z-2}$$

$$= z^2 Y(z) - 2Y(z)$$

$$\downarrow \text{I} \cdot z \cdot T$$

$$x[n] = y[n+2] - 2y[n]$$

$$Y(z) = \frac{z}{z-2}$$

$$\downarrow \text{I} \cdot z \cdot T$$

$$y[n] = 2^n [-u[-n-1]]$$

3: Exp multiplication (or) Scaling in z-domain:

For D.T.F.T we have

$$x[n] e^{j\omega_0 n} \longleftrightarrow X^* [e^{j(\omega - \omega_0)}]$$

for z.T  
Now

$$a = e^{j\omega_0} \quad \& \quad z = e^{j\omega}$$

$$a^n x[n] \longleftrightarrow X(z/a) \quad ; \quad \text{ROC} = |a|R$$

Application of the above property.

Ex:

$$y[n] = \cos \omega_0 n u[n]$$

$$= \left[ \frac{e^{j\omega_0 n} + e^{-j\omega_0 n}}{2} \right] x[n]$$

$$\text{let } x[n] = u[n]$$

$$X(z) = \frac{1}{1-z^{-1}} \quad ; \quad |z| > 1$$

$$Y(z) = \frac{X(z/e^{j\omega_0}) + X(z/e^{-j\omega_0})}{2}$$

$$Y(z) = \frac{1}{2} \left[ \frac{1}{1-z^{-1}e^{j\omega_0}} + \frac{1}{1-z^{-1}e^{-j\omega_0}} \right]$$

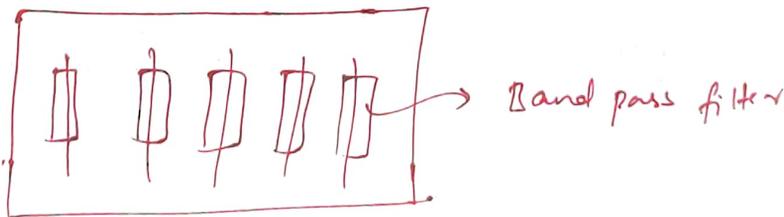
$$Y(z) = \frac{1 - z^{-1} \cos \omega_0}{1 - 2z^{-1} \cos \omega_0 + z^{-2}} \quad ; \quad |z| > 1 \cdot \left| e^{\pm j\omega_0} \right| > 1$$



Digital Resonator

Application: (Refer practical approach to D.S.P: Iffactor & Jarvis (LPE))

1. It is used as a 2<sup>nd</sup> order band pass filters. in 5-band Graphic Equaliser.



2. Used as 2<sup>nd</sup> order band pass filter in Touch tone DTMF.

Find  
is  
V(z)  
in the given  
matrix.

	1994	1995
(672)	1	2
(772)	4	5
	7	8

Knowledge of  $Y(z)$  is the basic formula for purchase of immediate  
 $\downarrow$  algorithm  
 D.F.T

Analysis of Digital S.P.  
 (By Ashok Ambekar)  
 Modern DSP

Echo  
 $\downarrow$   
 Towork filter  
 Used to reduce the echo

Download free ebooks (standard text books).

- (i) esnips.com
- (ii) ebook29.blogspot.com
- (iii) ebookee.com

$\rightarrow$  for digital communication refer  $\rightarrow$  (Gaussian waves.com)

7.2.7 sol  $X(z) = \frac{(z+j)(z-j)}{(z - \frac{1}{2})}$

$Y[n] = (\frac{1}{2})^n x[n] \Rightarrow Y(z) = X(\frac{z}{\frac{1}{2}})$   
 $(a = \frac{1}{2}) \quad = X(2z) = \frac{4z^2 + 1}{2z - \frac{1}{2}}$

## Time-reversal:

Reflection in the time domain is inversion in the  $z$ -domain.

$$x[-n] \longleftrightarrow X(z^{-1}) \quad ; \quad \text{ROC} = \frac{1}{R}$$

Eg:

$$(*) \quad y[n] = 3^n u[-n].$$

Sol:  $y[n] = \left[\frac{1}{3}\right]^{-n} u[-n]$  where  $x[n] = \left[\frac{1}{3}\right]^n u[n]$ .

$$= x[-n]$$

$$\downarrow z \cdot T$$

$$Y(z) = X(z^{-1})$$

$$Y(z) = \frac{1}{1 - \frac{z}{3}} \quad ; \quad |z| < \frac{1}{(1/3)}$$
$$\quad \quad \quad ; \quad |z| < 3$$

$$X(z) = \frac{1}{1 - \frac{1}{3}z^{-1}} \quad ; \quad |z| > \frac{1}{3}$$

---

## Differentiation in $z$ -domain:

$$n x[n] \longleftrightarrow -z \frac{d}{dz} X(z) \quad ; \quad \text{ROC} = R$$

for reference:  $-jn \longleftrightarrow \frac{d}{dw}$

$$z = e^{jw}$$

$$dz = j e^{jw} dw = j z dw$$

$$\boxed{dw = \frac{dz}{jz}}$$