

DFT:

→ N points (in one domain) $\xleftrightarrow{\text{transformed}}$ N other points (in other domain)

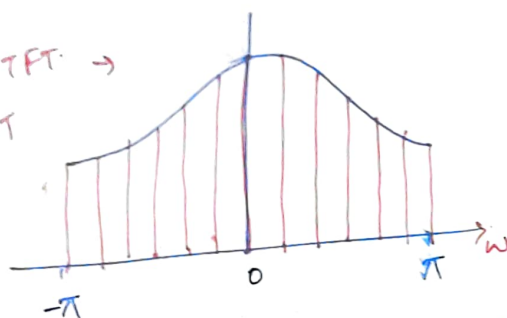
Periodic & Discrete \longleftrightarrow discrete & periodic

→ The sampled spectrum of DTFT is DFT.

→ So we will sample the DTFT which is known as DFT.

For DTFT $\rightarrow X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n}$

Sampled DTFT \rightarrow
i.e. DFT



$2\pi \rightarrow N$ points

$$\omega_k = \frac{2\pi k}{N}; k=0 \text{ to } (N-1)$$

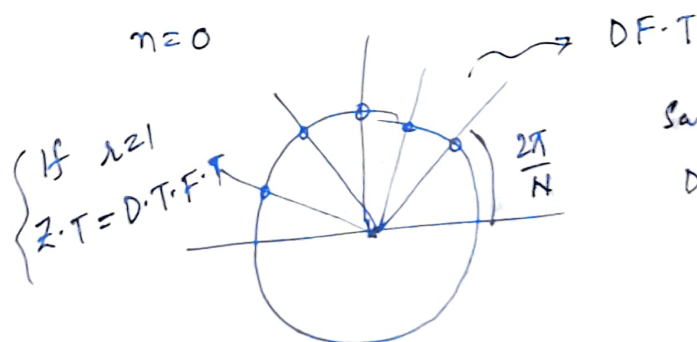
$\Delta f = \frac{2\pi}{N} = \text{frequency spacing (or) frequency resolution.}$

→ Now for DFT.

$$\Delta f = \frac{2\pi}{N} = \frac{f_s}{N}$$

$$X(e^{j\omega_k}) = \sum_{n=0}^{N-1} x[n] e^{-j\frac{2\pi}{N} \cdot kn}; k=0 \text{ to } (N-1)$$

$X[k]$



Sampled version of D.T.F.T is DFT.

→ Z-transform calculated on the unit circle at equidistant points i.e., an arc length of $\frac{2\pi}{N}$ is called as DFT.

→

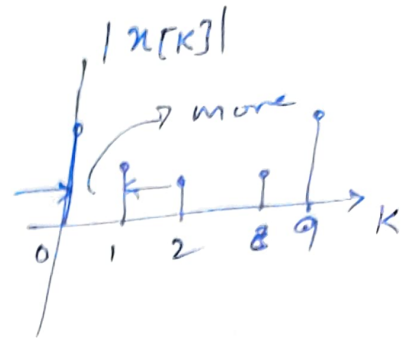
2π

$$\Delta f = \frac{2\pi}{N} = \frac{f_s}{N}$$

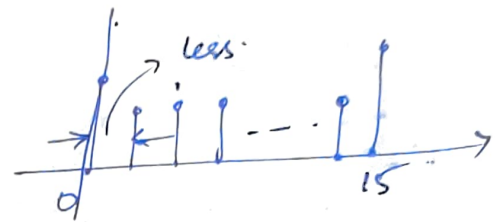
→ By adding extra zeros (zero padding) better frequency resolution is observed i.e., as the number of samples are increased the discrete spectrum will appear like continuous spectrum.

→ $N = 2^m$ pt.
 5pt $\Rightarrow 2^3$
 10pt $\Rightarrow 2^4$

Zero padding
 (↑ frequency resolution)



→ Phase factor $\omega_N = e^{-j2\pi/N}$



Properties of phase factor:

1. Periodicity: $\omega_{N+N} = \omega_N$

Proof: $(e^{-j\frac{2\pi}{N}})^{k+N} = e^{-j\frac{2\pi}{N}k} \cdot e^{-j2\pi} = \omega_N^k$

($\because \omega_4^6 = \omega_4^2$)

2. Symmetry: $\omega_{N+\frac{N}{2}} = -\omega_N$

3. $\omega_N^2 = \omega_{N/2}$

\Downarrow
 $(e^{-j\frac{2\pi}{N}})^2 = e^{-j\frac{4\pi}{N}} = e^{-j\frac{2\pi}{N/2}} = \omega_{N/2}$

→ Always D.F.T is calculated in terms of ω_N but not exponential terms.

→ For a particular function ω_N is fixed.

Now for DFT : $X[k] = \sum_{n=0}^{N-1} x[n] W_N^{kn}$; $k=0$ to $(N-1)$.

Matrix form:

$$\begin{bmatrix} X[0] \\ X[1] \\ X[2] \\ \vdots \\ X[N-1] \end{bmatrix} = \begin{bmatrix} 1 & 1 & 1 & \dots & 1 \\ 1 & W_N^1 & W_N^2 & \dots & W_N^{N-1} \\ 1 & W_N^2 & W_N^4 & \dots & W_N^{2(N-1)} \\ 1 & W_N^3 & W_N^6 & \dots & W_N^{3(N-1)} \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ 1 & W_N^{N-1} & W_N^{2(N-1)} & \dots & W_N^{(N-1)(N-1)} \end{bmatrix} \begin{bmatrix} x[0] \\ x[1] \\ x[2] \\ \vdots \\ x[N-1] \end{bmatrix}$$

for $N=2$: $W_2 = e^{-j2\pi/2} = -1$

$$W = \begin{bmatrix} 1 & 1 \\ 1 & W_2^1 \end{bmatrix} = \begin{bmatrix} 1 & 1 \\ 1 & -1 \end{bmatrix} \quad \text{ref}$$

for $N=4$: $W_4 = e^{-j2\pi/4} = e^{-j\pi/2} = -j$

$$W = \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & W_4^1 & W_4^2 & W_4^3 \\ 1 & W_4^2 & W_4^4 & W_4^6 \\ 1 & W_4^3 & W_4^6 & W_4^9 \end{bmatrix} = \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & W_4^1 & W_4^2 & W_4^3 \\ 1 & W_4^2 & W_4^4 & W_4^6 \\ 1 & W_4^3 & W_4^6 & W_4^9 \end{bmatrix}$$

Since $W_N^{k+N} = W$

$$= \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & -j & -1 & j \\ 1 & -1 & 1 & -1 \\ 1 & j & -1 & -j \end{bmatrix} \quad \text{ref}$$